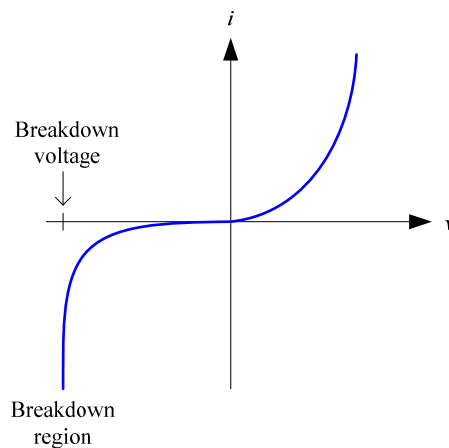


Lecture 6: Zener Diodes.

The very steep portion in the reverse biased i - v characteristic curve is called the **breakdown region**.



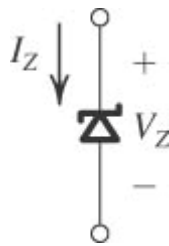
In this region the voltage across the diode remains nearly constant while the current varies (i.e., small internal resistance).

There are **two physical mechanisms** that can produce this behavior in the breakdown region. One is the **Zener effect** in which the large electric field in the depletion region causes electrons to be removed from the covalent bonds in the silicon.

The second mechanism is the **avalanche effect** in which charges that are accelerated to high speeds due to the large electric field in the depletion region collide with atoms in the silicon lattice causing charges to be dislodged. In turn, these dislodged charges have sufficient energy to liberate additional electrons. In other words, this avalanche effect is a cascading, ionization process.

Provided that the power dissipated in the diode is less than the maximum rated, the diode is **not damaged** when operating in the breakdown region. In fact, Zener diodes are designed to operate in this region.

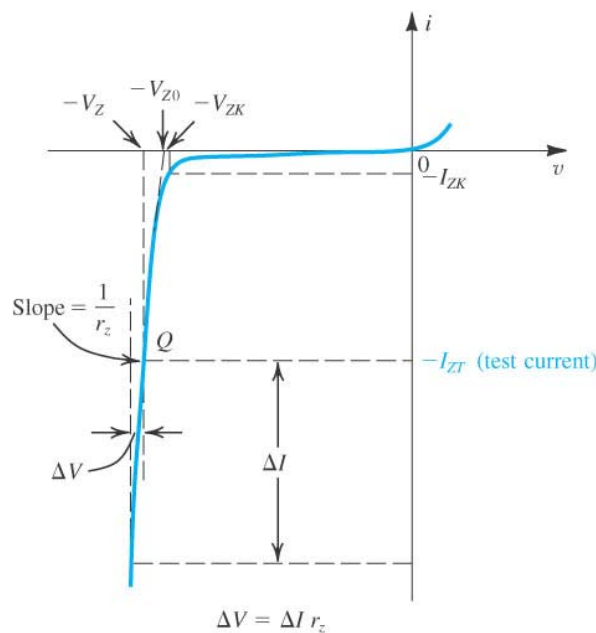
The circuit symbol for the Zener diode is



(Fig. 3.20)

These diodes are usually operated in the reverse bias regime (i.e., breakdown region) so that $I_Z > 0$ and $V_Z > 0$.

An enlargement of this breakdown region is shown in text Figure 3.21:

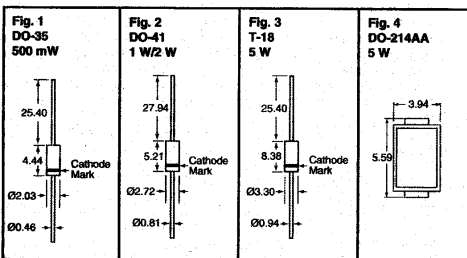


(Fig. 3.21)

The manufacturer specifies the $-V_{Z0}$ and test current I_{ZT} . One can design Zeners with a wide range of voltages.

The page below is from a Digikey catalog (www.digikey.com) and shows voltages from ranging from 3.6 V to 200 V, for example.

Microsemi. Zener Diodes



500 mW Series DO-35 Silicon Zener Diodes, 1N5221 — 1N5281

Nominal Zener Voltage V_Z @ I_{ZT} (Volts)	Test Current I_{ZT} (mA)	Nominal Zener Voltage V_Z @ I_{ZT} (Volts)	Test Current I_{ZT} (mA)	Nominal Zener Voltage V_Z @ I_{ZT} (Volts)	Test Current I_{ZT} (mA)	Nominal Zener Voltage V_Z @ I_{ZT} (Volts)	Test Current I_{ZT} (mA)	Nominal Zener Voltage V_Z @ I_{ZT} (Volts)	Test Current I_{ZT} (mA)
21	2.4	35	6.8	46	16	60	43	71	100
23	2.7	36	7.5	47	17	61	47	72	110
24	2.8	37	8.2	48	18	62	51	73	120
26	3.3	38	8.7	49	19	63	56	74	130
27	3.6	39	9.1	50	20	64	60	75	140
28	3.9	40	10	51	22	65	62	76	150
29	4.3	41	11	52	24	66	68	77	160
30	4.7	42	12	53	25	67	75	78	170
31	5.1	43	13	54	27	68	82	79	180
33	6.0	44	14	55	28	69	87	81	200
34	6.2	45	15	58	36	70	91		

1W Series, DO-41 Glass Zener Diodes
1N4729A — 1N4763A

Zener Voltage V_Z (Volts)	Test Current I_{ZT} (mA)	Zener Voltage V_Z (Volts)	Test Current I_{ZT} (mA)
29	3.6	41	11
33	5.1	42	12
34	5.6	43	13
35	6.2	44	14
36	6.8	45	15
37	7.5	46	16
38	8.2	47	17
40	10	48	18

2 W Series, DO-41 Silicon Zener Diodes
2EZ3.6D5 — 2EZ200D5

Nominal Zener Voltage V_Z (Volts)	Test Current I_{ZT} (mA)	Nominal Zener Voltage V_Z (Volts)	Test Current I_{ZT} (mA)	Nominal Zener Voltage V_Z (Volts)	Test Current I_{ZT} (mA)
3.6	3.6	17	17	62	8.1
3.9	3.9	18	18	75	6.7
4.3	4.3	19	19	82	6.1
5.1	5.1	20	20	91	5.5
6.2	6.2	22	22	110	4.5
6.8	6.8	27	27	120	4.2
7.5	7.5	33	33	130	3.8
8.2	8.2	39	39	140	3.6
12	12	43	43	170	2.9
13	13	47	47	180	2.8
14	14	51	51	190	2.6
15	15	56	56	200	2.5

5 W Series, T-18 Silicon Zener Diodes
1N5333B — 1N5388B

Regulator Voltage V_Z (Volts)	Test Current I_{ZT} (mA)	Regulator Voltage V_Z (Volts)	Test Current I_{ZT} (mA)
33	3.3	58	22
34	3.6	65	30
36	4.3	68	25
39	5.6	70	20
40	6.0	73	20
42	6.8	75	15
43	7.5	76	15
47	10	79	12
49	12	81	10
51	14	86	5.0
52	15	87	5.0
53	16	88	5.0

Fig.	DC Power Dissipation	Voltage Range (Code) See Chart	Digi-Key Part No.*	1	10	100	Digi-Key Part No.*	2,500	3,000	Tape and Reel Pricing 4,000	6,000	10,000	Microsemi Part No.*
1	500 mW	2.4V — 200V	1N52(Value)DO35MSCT-ND	1.20	8.40	60.00	1N52(Value)DO35MSTR-ND	—	—	—	—	210.00/M	1N52(Value)DO35
2	1 W	3.6V — 91V	1N47(Value)DO41MSCT-ND	.25	1.65	13.20	1N47(Value)DO41MSTR-ND	—	—	74.00/M	—	59.00/M	1N47(Value)DO41
	2 W	3.6V — 200V	2EZ(Value)D50041MSCT-ND	1.42	9.45	75.60	2EZ(Value)D50041MSTR-ND	—	—	293.00/M	—	340.00/M	2EZ(Value)D50041
3	5 W	3.3V — 200V	1N53(Value)BM5CT-ND	1.20	8.00	64.00	1N53(Value)BM5STR-ND	—	400.00/M	—	320.00/M	—	1N53(Value)B

* For complete part number, substitute corresponding value from chart for (Value).

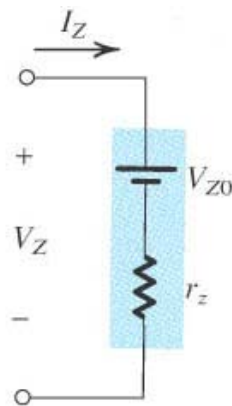
The rated V_Z at the specified I_{ZT} is listed for these Zener diodes. The circled component, for example, has $V_Z = 8.2$ V at $I_{ZT} = 31$ mA. The maximum rated power is 1 W for this device.

As the current deviates from the specified value I_{ZT} , the voltage V_Z also changes, though perhaps only by a small amount. The change in voltage ΔV_Z is related to the change in the current ΔI_Z as

$$\Delta V_Z = r_z \Delta I_Z \quad (1)$$

where r_z is the **incremental or dynamic resistance** at the Q point and is usually a few Ohms to tens of Ohms. See the datasheet for the particular device you are working with.

Because of the nearly linear relationships in the breakdown region, the reverse bias model of the Zener diode is



(Fig. 3.22)

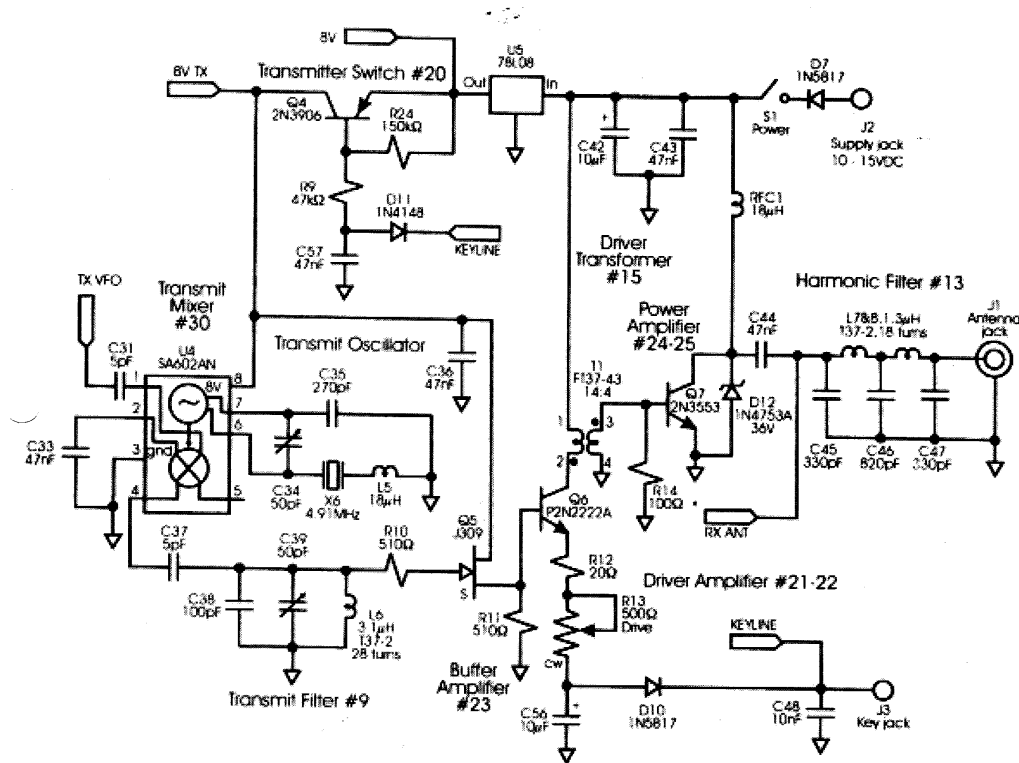
where $V_Z = V_{Z0} + r_z I_Z$ (3.20),(2)

as is apparent from Fig. 3.21.

Applications of Zener Diodes

What are Zener diodes used for? Applications include:

1. **Voltage overload protection.** This circuit is from the NorCal 40A radio that is built in EE 322 *Electronics II – Wireless Communication Electronics*:



2. **Voltage regulation.** See the figure below. An example of such a regulator circuit will be considered next.

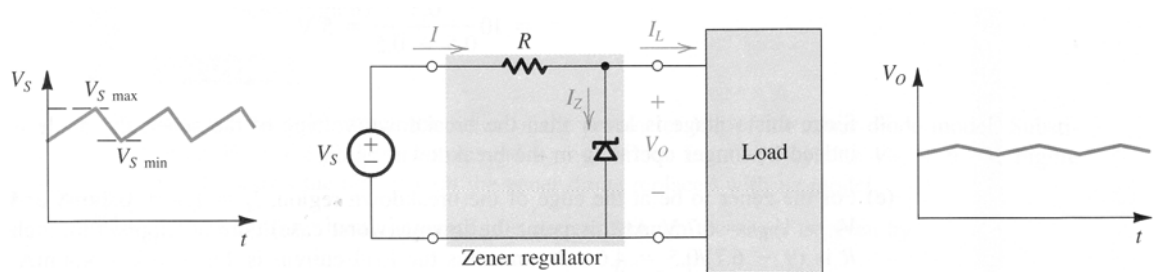
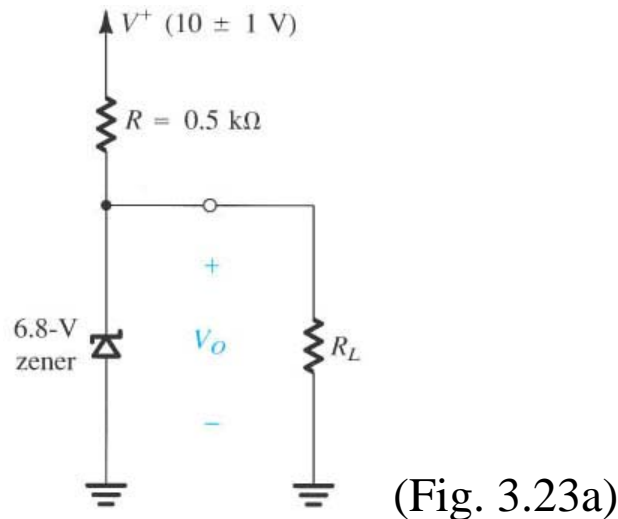


Fig. 3.34 A zener shunt regulator. Observe that while the raw supply V_S has a large ripple component, the regulated voltage V_O has a very small ripple.

(Source: Sedra and Smith, fourth ed.)

Example N6.1 (similar to text example 3.8). The Zener diode in the circuit below has the following characteristics: 6.8-V rating at 5 mA, $r_z = 20\ \Omega$, and $I_{ZK} = 0.2\text{ mA}$.



With these ratings

$$V_Z = V_{Z0} + r_z I_Z \Rightarrow V_{Z0} = V_Z - r_z I_Z$$

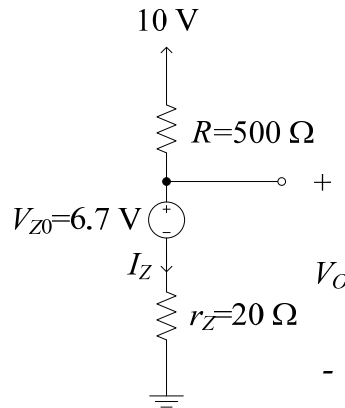
or

$$V_{Z0} = 6.8 - 20 \cdot 5 \times 10^{-3} = 6.7\text{ V}$$

Note that the supply voltage can fluctuate by $\pm 1\text{ V}$. Imagine this fluctuation is a random process rather than a time periodic variation.

Determine the following quantities:

- (b) Find V_O with no load and V^+ at the nominal value. The equivalent circuit for the reverse bias operation of the Zener diode is



From this circuit we calculate

$$I_Z = \frac{10 - 6.7}{500 + 20} = 6.35 \text{ mA}$$

Therefore,

$$V_O = 10 - I_Z \cdot 500 = 10 - 6.35 \times 10^{-3} \cdot 500 = 6.83 \text{ V}$$

- (c) Find the change in V_O resulting from a $\pm 1 \text{ V}$ change in V^+ .
Using the circuit above the $V^+ = 11 \text{ V}$:

$$V_O = 11 - \frac{11 - 6.7}{500 + 20} \cdot 500 = 6.865 \text{ V}$$

Similarly, with $V^+ = 9 \text{ V}$:

$$V_O = 9 - \frac{9 - 6.7}{500 + 20} \cdot 500 = 6.788 \text{ V}$$

Consequently, $\Delta V_O = 6.865 - 6.788 = 0.077 \text{ V}$ or
 $\Delta V_O = \pm 38.5 \text{ mV}$.

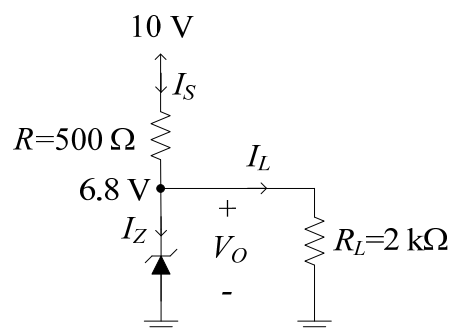
The ratio of the change in output voltage to the change in the source voltage ($\Delta V_O / \Delta V^+$) is called the **line regulation**

of the regulator circuit. It's often expressed in units of mV/V. For this example and no load attached,

$$\text{Line Regulation} \equiv \frac{\Delta V_o}{\Delta V^+} = \frac{77}{11-9} \frac{\text{mV}}{\text{V}} = 38.5 \frac{\text{mV}}{\text{V}}$$

- (d) Find the change in V_o resulting from connecting a load of $R_L = 2 \text{ k}\Omega$ with a nominal $V^+ = 10 \text{ V}$.

Assuming that the diode is operating in the breakdown region:



then
$$I_L = \frac{6.8}{2000} = 3.4 \text{ mA}.$$

Is this a reasonable value? Calculate I_S :

$$I_S = \frac{10 - 6.8}{500} = 6.4 \text{ mA}.$$

So, **yes, this is a reasonable** value because $I_L < I_S$, as it must.

From (1), $\Delta V_o = r_z \Delta I_Z$ and since $\Delta I_Z = -3.4 \text{ mA}$ then

$$\Delta V_o = 20(-3.4 \times 10^{-3}) = -68 \text{ mV}$$

The ratio of the change in output voltage to the change in the load current ($\Delta V_o / \Delta I_L$) is called the **load regulation** of the regulator circuit. It's often expressed in units of mV/mA. For this example,

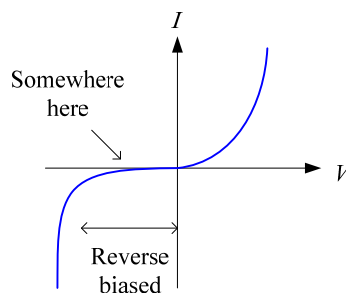
$$\text{Load Regulation} \equiv \frac{\Delta V_o}{\Delta I_L} = \frac{77 \text{ mV}}{-3.4 \text{ mA}} = -22.6 \frac{\text{mV}}{\text{mA}}$$

- (e) What is V_o when $R_L = 0.5 \text{ k}\Omega$? Assume the diode is in breakdown. In this case,

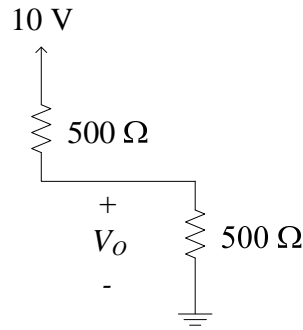
$$I_L \approx \frac{6.8}{500} = 13.6 \text{ mA}.$$

Is this a reasonable value? **No**, because this value is greater than $I_S = 6.4 \text{ mA}$.

Therefore, in this case the Zener diode is not operating in the breakdown region. Also, the diode can't be forward biased. Consequently, we conclude the diode must be operating in the **reverse bias region**.



The equivalent circuit in this case is



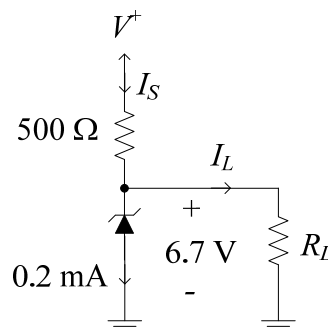
From this circuit we calculate

$$V_o = \frac{500}{500 + 500} \cdot 10 = 5 \text{ V}.$$

This voltage is less than the breakdown voltage V_{ZK} , which is **consistent** with the reverse biased assumption.

- (f) Determine the minimum R_L for which the diode still remains in breakdown for all V^+ . (We know from the results in parts (c) and (d) of this example that R_L must lie between 500Ω and $2 \text{ k}\Omega$ when $V^+ = 10 \text{ V}$.)

Referring to Fig. 3.21, at the “knee” $I_Z = I_{ZK} = 0.2 \text{ mA}$ and $V_Z = V_{ZK} \approx V_{Z0} = 6.7 \text{ V}$.



- If $V^+ = 9$ V:

$$I_S = \frac{9 - 6.7}{500} = 4.6 \text{ mA.}$$

Therefore, $I_L = 4.6 \text{ mA} - 0.2 \text{ mA} = 4.4 \text{ mA}$, so that

$$R_L = \frac{V_L}{I_L} = \frac{6.7}{4.4 \times 10^{-3}} = 1,522 \text{ } \Omega$$

- If $V^+ = 11$ V:

$$I_S = \frac{11 - 6.7}{500} = 8.6 \text{ mA.}$$

Therefore, $I_L = 8.6 \text{ mA} - 0.2 \text{ mA} = 8.4 \text{ mA}$, so that

$$R_L = \frac{V_L}{I_L} = \frac{6.7}{8.4 \times 10^{-3}} = 798 \text{ } \Omega$$

The **smallest load resistance** that can be attached to this circuit and have the diode remain in breakdown is $R_L = 1,522 \text{ } \Omega$. The reason is that for any smaller value when $V^+ = 9$ V results in the diode leaving breakdown and entering the reverse bias mode.